

BLANKING NOISE

Consider the total power output voltage of a continuum microwave radiometer to be described statistically as a stationary gaussian random variable. The variance of the voltage $v(t)$ is defined as

$$V[v(t)] \equiv C^2 [\Delta T_{\text{RMS}}]^2 \quad (1)$$

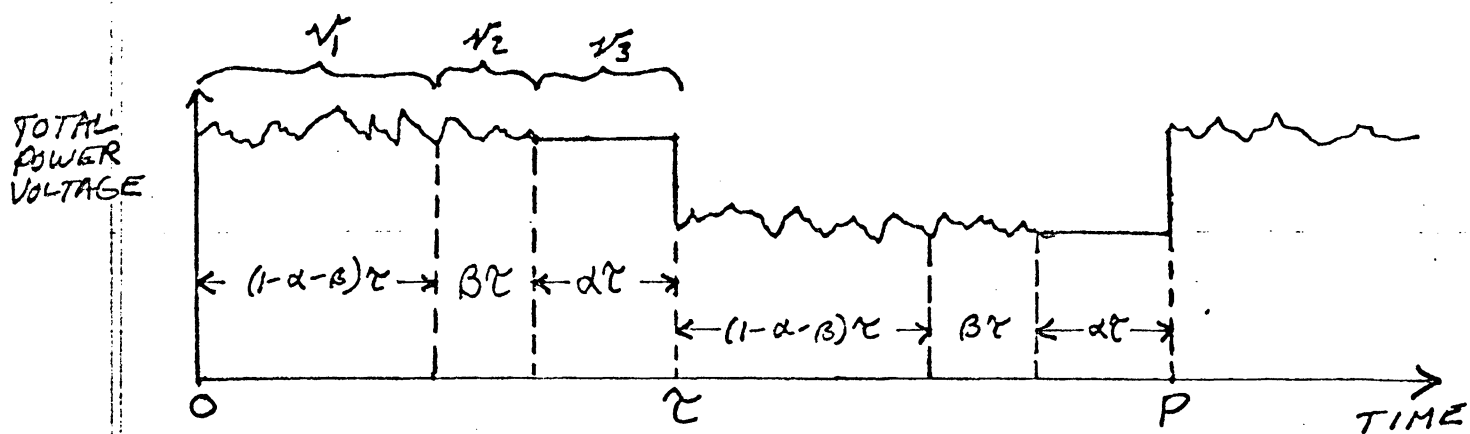
for one second of integration time per sample. Neglecting gain variations, one can calculate the theoretical noise referred to the input power spectral density as

$$\Delta T_{\text{RMS}} = \frac{KT_{\text{SYSTEM}}}{\sqrt{\text{Bandwidth}}} \quad (2)$$

where $K=1$ for a total power radiometer.

Suppose we blank the total power signal for a relative fraction α of the time in one half-cycle of a Dicke switch period and then synchronously detect the signal. How much will the output noise change?

Let $P \equiv$ Dicke switch period
 $\tau \equiv P/2 =$ Time during one half-cycle
 $\alpha\tau \equiv$ Time the signal is held constant by the blanking circuit
 $\beta\tau \equiv$ Time the signal is integrated to determine the holding voltage



The parameters ν_i represent the mean value of $v(t)$ during the sections of the half-cycle as indicated in the drawing. Formally,

$$\nu_1 \equiv \frac{1}{(1-\alpha-\beta)\tau} \int_0^{(1-\alpha-\beta)\tau} v(t) dt \quad (3)$$

$$\nu_2 \equiv \frac{1}{\beta\tau} \int_{(1-\alpha-\beta)\tau}^{(1-\alpha)\tau} v(t) dt \quad (4)$$

$$\nu_3 \equiv \frac{1}{\alpha\tau} \int_{(1-\alpha)\tau}^{\tau} v(t) dt \quad (5)$$

The mean value of $v(t)$ averaged over the entire half-cycle is defined as ν_0 and is given by

$$\nu_0 \equiv \frac{1}{\tau} \int_0^{\tau} v(t) dt, \quad (6)$$

$$\text{and } \therefore \nu_0 = (1-\alpha-\beta)\nu_1 + \beta\nu_2 + \alpha\nu_3. \quad (7)$$

Recall that the holding voltage ν_3 is simply the mean of $v(t)$ during the period of length $\beta\tau$ immediately preceding the holding time. That is, $\nu_3 = \nu_2$.

Therefore,

$$v_0 = (1-\alpha-\beta)v_1 + (\alpha+\beta)v_2 \quad (8)$$

The variance of v_0 is

$$V[v_0] = (1-\alpha-\beta)^2 V[v_1] + (\alpha+\beta)^2 V[v_2] \quad (9)$$

where v_1 and v_2 are clearly independent variables and their correlation coefficient is zero. The variances are simply given by

$$V[v_1] = \left[\frac{C \cdot \Delta T_{RMS}}{\sqrt{(1-\alpha-\beta)} \tau} \right]^2 \quad (10)$$

$$\text{and } V[v_2] = \left[\frac{C \cdot \Delta T_{RMS}}{\sqrt{\beta} \tau} \right]^2 \quad (11)$$

$$\therefore V[v_0] = \frac{(1-\alpha-\beta)^2 C^2 \Delta T_{RMS}^2}{(1-\alpha-\beta) \tau} + \frac{(\alpha+\beta)^2 C^2 \Delta T_{RMS}^2}{\beta \tau} \quad (12)$$

$$\therefore V[v_0] = \left[\frac{C \Delta T_{RMS}}{\sqrt{\tau}} \right]^2 \left[1-\alpha-\beta + \frac{\alpha^2+2\alpha\beta+\beta^2}{\beta} \right] \quad (13)$$

$$\text{Define } \Delta_0 \equiv \sqrt{V[v_0]} = \frac{C \Delta T_{RMS}}{\sqrt{\tau}} \sqrt{1+\alpha+\alpha^2/\beta} \quad (14)$$

In the absence of blanking, the standard deviation of the mean voltage during a half-cycle is

$$\Delta_{NB} \equiv \frac{C \Delta T_{RMS}}{\sqrt{\tau}} \quad (15)$$

and the factor by which blanking increases the RMS noise out of the synchronous detector is

$$F \equiv \frac{\Delta_0}{\Delta_{NB}} = \sqrt{1+\alpha+\alpha^2/\beta} \quad 16$$

It is immediately obvious that blanking can only increase the noise level. This is strictly true only if $\beta \leq 1 - \alpha$, for which the preceding analysis is valid. Values of $\beta > 1 - \alpha$ are not allowed since information will be lost. That is, $v(t)$ will be integrated over both signal and reference halves of the full period, and the difference output of the synchronous detector will be reduced. Smoothing of alternate half-cycles is known as "reference averaging" and is useful for load switching, but not for beam switching. Only in this case can F be less than one.

"Optimum" blanking occurs when $\beta = 1 - \alpha$,

$$F_{opt} \equiv F|_{\beta=1-\alpha} = \sqrt{1 + \alpha + \frac{\alpha^2}{1-\alpha}} \quad (17)$$

$$F_{opt} = \frac{1}{\sqrt{1-\alpha}} \quad (18)$$

That is, in the optimum case, the noise is just what one would get with no blanking and a half-cycle of length $(1-\alpha)\tau$. Clearly, the only effect is reduced integration time.

CASE I: NRAO SBE: $\alpha = 0.32$, optimum case

$$F_{opt} = \frac{1}{\sqrt{1-0.32}} = \underline{1.21}$$

$$\text{Measured } F = 1.22 \pm 0.05$$

CASE II : $\alpha = 0.15$, $\beta = 0.33$

S/R Freq. = 1.25 Hz
 $\tau = 400$ ms
60 ms blanking
RC = 66 ms on input
to sample/hold

$$F = \sqrt{1 + 0.15 + 0.15^2/0.33} = 1.10 \leftarrow$$

$$F(\text{Measured}) = 1.12 \pm 0.08 \leftarrow$$

CASE III : $\alpha = 0.3$, $\beta = 0.04$

$$F = \sqrt{1 + 0.3 + 0.3^2/0.04} = 1.9 \leftarrow$$

$$F(\text{Measured}) = 1.7 \pm 0.2 \leftarrow$$

NOTE :

When calculating α and β , remember that the equivalent integration time of an RC filter is 2RC.

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